# SURROUND SOUND PROCESSOR

Based on a design by D. Laues

The processor described in this article expands the sound of an existing stereo TV receiver or audio installation with a centre channel and a surround channel. It does not use special Dolby ICs. Additional output amplifiers are not needed, because they are provided in the processor.

reating surround sound in a domestic room can be approached in two ways: by a processor that generates the four signals required: left-hand, right-hand, centre and surround, or by one that adds the two missing channels, that is, centre and surround, to the existing stereo sound. The first is the most elegant, but also the most complicated and most expensive. Moreover, it requires an additional line to return the left-hand and right-hand signals to the input of the TV/audio equipment. The second way is much more straightforward and has proved in practice to give an excellent spatial effect. Moreover, it can be accomplished in a compact and fairly inexpensive unit. The design in this article is of the second kind.

The basic setup in a domestic room is shown in Fig. 1. The left-hand and righthand channels are reproduced as before, that is, via the loudspeakers in the TV receiver, as shown, or by those of the audio installation to which the TV receiver is connected. The extra items are the processor and three loudspeakers. The inputs of the processor are linked, possibly via

the SCART connector, with the line out terminal of the TV receiver, or audio amplifier, while the extra loudspeakers are connected to the outputs of the processor. The processor contains two integral amplifiers each of which provides 20 W output into 4  $\Omega$ : quite sufficient for the centre and surround loudspeakers.

As briefly discussed in last month's article on Dolby surround sound, the additional loudspeakers need be no more than compact (bookcase type) hi-fi types that are not too expensive. If possible, however, choose types whose efficiency is about the same as that of the main loudspeakers: this affords rather more freedom when the system is set up as a whole.

# **Basic design**

A block diagram of the basic design is given in Fig. 2. The design of the processor is similar to that of the active decoder discussed in last month's article on Dolby surround sound. There are some differences between the two, since the present processor not only functions as a decoder, but also provides the signals for the centre and surround channels and contains two output amplifiers.

As explained last month, the two additional channels are processed (whence the name of the unit) from the sum and difference of the two stereo channels. The centre channel is the simpler to produce, because it suffices to add the left-hand and right-hand channels together and apply the sum to a suitable output amplifier via a voltage-controlled amplifier (VCA).

To produce the surround signal, the righthand signal is deducted from the left-hand signal (or vice versa) and the resulting signal is applied to a delay network via a low-pass filter. The delay can be preset between 10 ms and 30 ms. The signal is then applied to an expander via a band-pass filter. The expander is essential because the surround signal is compressed during recording. The output of the expander is applied to a second output amplifier via a VCA.

The dynamic compensating network, in conjunction with the VCAs, reflects the dif-

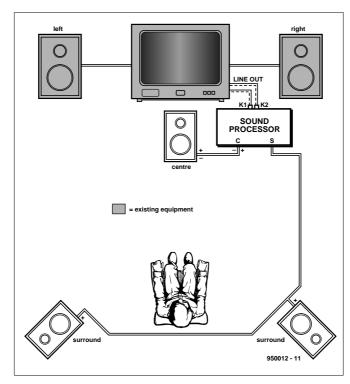


Fig. 1. Basic setup of a surround sound system in a living room.



ference between an active and a passive (matrix) design. In this network, the correlation between the two stereo channels is analysed continuously. The results of the analysis are converted into control signals for the VCAs which constantly adjust the levels of the centre and surround signals. This arrangement ensures a much larger channel separation than possible with a passive design.

## Centre channel

From the inputs of the left-hand and righthand channels,  $K_1$  and  $K_2$  in Fig. 5, the signals are taken via buffer amplifiers  $IC_{1a}$  and  $IC_{1b}$  to  $R_{51}$  and  $R_{52}$ , and then summed in preset  $P_6$ . From there, the signal is applied to pin 5 of  $IC_9$ . This IC contains four electronic potentiometer circuits (of which only two are used) which function as VCAs; their amplification is governed by a control voltage at pins 9 and 10. The output of one of the circuits is available at pin 7, from where it is applied to

output amplifier IC<sub>10</sub>. This circuit provides an output of up to 20 W into 4  $\Omega$ .

The output of  $IC_{10}$  is applied to the centre channel loudspeaker via relay contact  $Re_{1b}$ . The relay is controlled by a simple delay circuit,  $T_1$ , and obviates any clicks and plops in the speaker caused by the switching on and off of the processor.

# Surround channel

The signals at the outputs of IC1a and IC<sub>1b</sub> are also applied to the inverting and non-inverting inputs of IC<sub>2b</sub> respectively, so that the output of the op amp is the difference of the two stereo signals (L-R). The difference signal is applied to a 4th-order lowpass filter based on  $IC_{2a}$ , which limits its upper bandwidth to 7 kHz. This anti-aliasing filtering serves to obviate the formation of spurious mixing products of the signal and the clock of the following delay line, which is based on IC4. This IC is a 2048-stage bucket brigade device. The rate at which the internal electronic switches are operated is

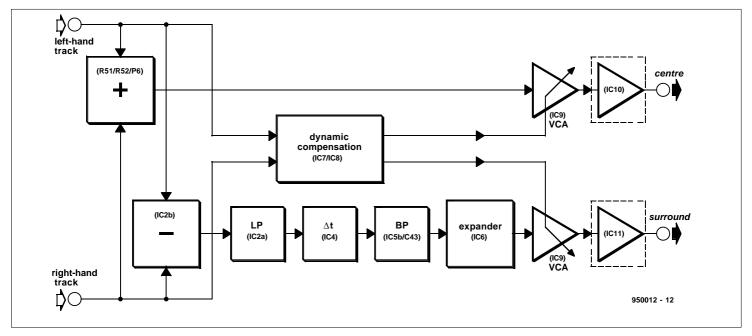


Fig. 2. Block diagram of the surround sound processor.

determined by IC<sub>3</sub>. This CMOS-IC is designed especially to generate a low-impedance, double-phase clock. The specified values of its frequency-determining components,  $R_{15}$ ,  $R_{16}$ ,  $C_9$  and  $P_1$  allow a delay between 10 ms and 30 ms to be set with  $P_1$ .

The outputs of IC<sub>4</sub> are applied via buffer IC<sub>5a</sub> to a low-pass filter based on IC<sub>5b</sub> (identical to that based on IC<sub>2a</sub>) which filters out any residue of the clock signal. The cut-off frequency is 7 kHz. The signal is subsequently fed to compander IC<sub>6</sub>, whose input network contains a high-pass filter,  $R_{55}$ -C<sub>43</sub>, the specified values of which give a lower cut-off frequency of about 50 Hz. The overall effect of the low-pass and high-pass filters is, of course, that of a band-pass filter as shown in Fig. 2.

The compander IC contains two circuits each consisting of a rectifier, a variable gain cell and an op amp. In the present processor only one of these circuit is used and that as an expander. The values of external components  $R_{27}$ - $R_{32}$  and  $C_{23}$ - $C_{29}$  allow for an expansion factor of 1:1.3.

The surround signal is then applied to the second electronic potentiometer circuit in  $IC_9$ , whose output is available at pin 17. From there, the signal is fed to output amplifier  $IC_{11}$ , whose amplification is identical to that of  $IC_{10}$ . The output of  $IC_{11}$  is applied to the surround loud-speaker(s) via a second contact on  $Re_1$ .

### Dynamic compensation

The outputs of buffers  $IC_{1a}$  and  $IC_{1b}$  are also applied to twin comparators  $IC_{7b}$  and  $IC_{7c}$  via  $C_{31}$  and  $C_{32}$ . The output of each of these comparators is a rectangular voltage the frequency of which is a measure of the variation in the relevant input signal. Both outputs are applied to XOR gate  $IC_{8c}$ . (Remember that an XOR gate has an output only when its inputs are dissimilar). Integration of the output pulses of the gate by  $R_{37}$ - $C_{36}$  results in a direct voltage whose amplitude is a measure of the phase difference between the two stereo signals.

This direct voltage is applied via  $IC_{7a}$  (inverted) and  $IC_{7d}$  (non-inverted) to the control inputs (pins 9 and 10) of  $IC_9$ . This arrangement ensures that when a mono signal is present at the inputs (no or hardly any phase difference), the amplification of the VCA controlling the centre channel is raised. Conversely, when a surround signal is present (large phase difference), the amplification of the VCA controlling the surround channel is increased.

The degree to which the amplification of the VCAs is influenced by the control signals is preset by  $P_2$  and  $P_3$ . When the wipers of these controls are at earth potential, the amplification is fixed; when they are at the opposite end of their travel, control is maximum.

Presets  $P_4$  and  $P_5$  serve to shift the operating point of the VCAs to some extent. They thus make the preset range wider and, in fact, support the operation of  $P_6$  and  $P_7$ . If, for instance,

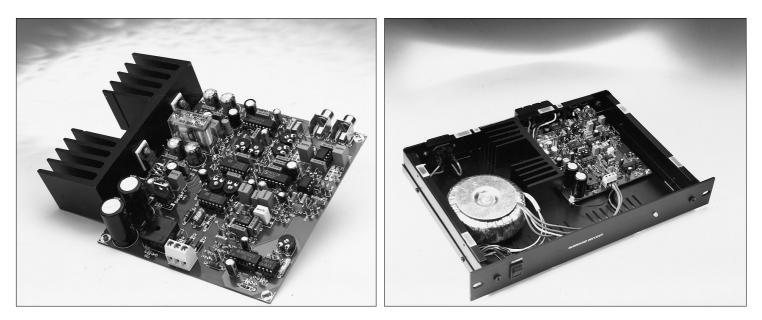
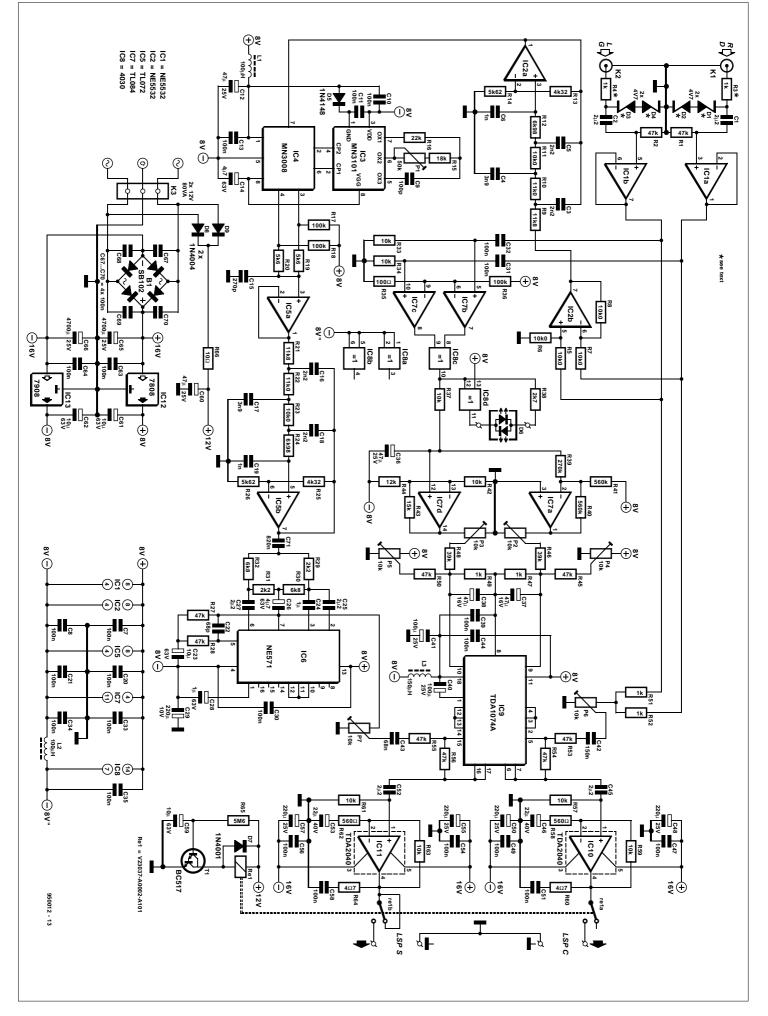


Fig. 3. Completed prototype board.

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 $P_7$  has already set the surround level to maximum,  $P_5$  enables this to be increased slightly. The same applies to  $P_4$  insofar as the level of the centre channel preset with  $P_6$  is concerned.

The currents through  $R_{45}$  and  $R_{46}$  and those through  $R_{48}$  and  $R_{50}$  are simply added together: there is, therefore, no interaction between  $P_2$  and  $P_4$  nor between  $P_3$  and  $P_5$ .

## Further circuit details

Resistors  $R_3$  and  $R_4$  and diodes  $D_1-D_4$  limit the level of the input signal to a safe value and are imperative if the stereo signals are taken from the loudspeaker outputs of the TV receiver. Note that even line out terminals sometimes provide a signal at a level well above 1 V. If it is absolutely certain that the line output level is 1 V, and this is the only input, the resistors can be replaced by a wire bridge and the diodes may be omitted.

Bi-colour LED D<sub>6</sub> functions as a kind of signal monitor that shows the change from surround channel to centre channel and vice versa. In the case of a surround signal, the output of  $IC_{8c}$  is high. Since one output of  $IC_{8d}$  is at + 8 V, both inputs of this XOR are then high, so that its output is low. This results in the red segment of D<sub>6</sub> lighting. In the case of a centre signal, pin 12 of IC<sub>8d</sub> is low, so that its output is high, resulting in the green segment of D<sub>6</sub> lighting. In practice, the changes between the signal are so rapid that the LED shows a fluent transition from red to green and back to red again.

The power supply provides three different voltages. The secondary of the mains transformer is connected to  $K_3$ . The 12 V input, after rectification and smoothing, results in a symmetrical supply of  $\pm$  16 V, which is used to power output amplifiers IC<sub>10</sub> and IC<sub>11</sub>.

From the  $\pm$  16 V lines, voltage regulators IC<sub>12</sub> and IC<sub>13</sub> derive a supply of  $\pm$  8 V, which is used to power the remainder of the circuit.

The 12 V line for the relay is taken directly from  $K_3$  and rectified by  $D_8$ - $D_9$ .

## Construction

The processor is best constructed on the PCB illustrated in Fig. 6. Since this board is double sided and through-plated, it is not possible to make it without special tools and equipment.

Populating the board is straight-

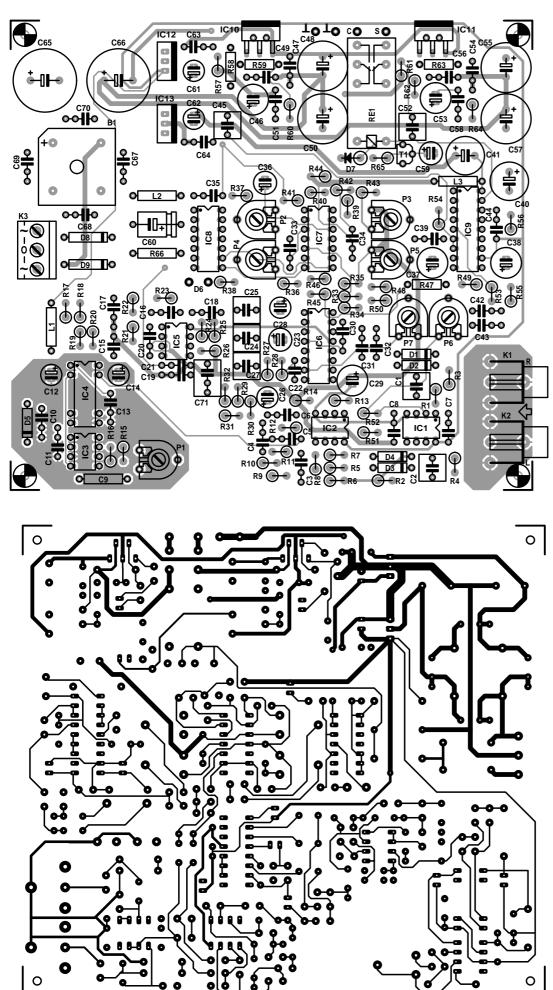


Fig. 6. Printed-circuit board for the surround sound processor. (See also next page)

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forward and should not present undue difficulties. Note that when the specified enclosure is used, capacitors  $C_{65}$  and  $C_{66}$  must be not higher than 38–40 mm.

Use gold-plated connectors for  $K_1$  and  $K_2$  to ensure good, lasting connections. The positions for these connectors are at the edge of the board so that all that is necessary when fitting the board into the enclosure is drilling a few holes in the back panel through which these sockets can protrude. Note that they must not touch the enclosure.

Output amplifiers  $IC_{10}$  and  $IC_{11}$ are located at the edge of the board so that they can be fitted readily to a heat sink. The ICs must be electrically isolated from the heat sink by ceramic washers and heat conducting paste.

The photograph in Fig. 3 shows the completed prototype board.

The main requirements of the enclosure are that it is made of metal and that it provides adequate space for the finished board.

Apart from  $K_1$  and  $K_2$ , fit suitable sockets or spring-loaded terminals at the back of the enclosure for connecting the centre and surround loudspeakers. Link these terminals with heavy-duty insulated

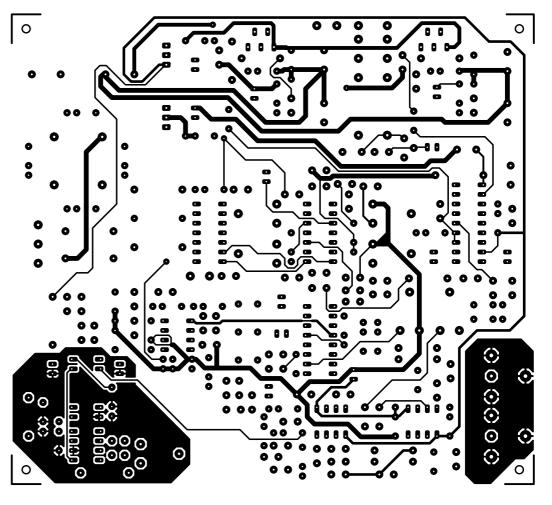
wire to points 'C' and 'S' and the adjacent earthing points on the board. The specified mains entry with integral fuse holder should also be fitted at the back of the enclosure.

The mains on/off switch and  $D_6$  should be fitted at the front panel (for which a readymade foil is not available). The diode should be connected to the relevant points on the board via lengths of flexible stranded wire. Finally, link the centre pin of K<sub>3</sub> to the mass

of the enclosure with the aid of a solder tag.

The completed prototype is shown in the photograph of Fig. 4.

The output amplifiers are suitable for operation with load impedances  $\geq 4 \Omega$ , but not



lower ones. Thus, for the surround channel, two 8  $\Omega$  loudspeakers may be connected in parallel only if it is absolutely certain that the impedance is 8  $\Omega$ . If it is not, connect the speakers in series: this is safer. It is essential that the two speakers are in phase: the + terminals must go to the same terminal on the board: whether this is earth or 'S' does not matter.

The centre loudspeaker must be in phase with the main speakers. Since the relevant VCA functions as an inverter, the + terminal of this speaker must be connected to the earth point on the board; the -terminal to point 'C'. Start by setting presets  $P_2$ - $P_7$  to the centre of their travel, and  $P_1$  to maximum (fully clock-wise).

Inject a (mono) speech signal and adjust  $P_6$ until the sound appears to come from the centre speaker. Wait for a surround signal (indicated by the red segment of  $D_6$  lighting) and turn  $P_7$  till sound emanates from the surround speakers. Do not set the level too high, because this leads quickly to an exaggerated effect. If, however, it is felt that the desired level can not be obtained with  $P_6$  or  $P_7$ , as the case may be, adjust  $P_4$  or  $P_5$ , or both, as required.

Next, create a spatial effect by slowly turning  $P_2$  and  $P_3$  until the centre channel and the

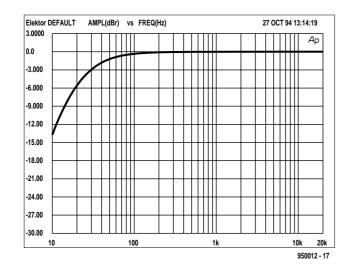


Fig. 7. Frequency response of the centre channel.



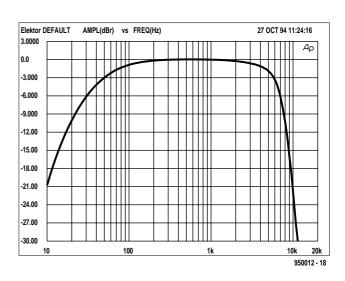


Fig. 8. Frequency response of the surround channel.

surround channel seem well 'separated'. It is more than likely that  $P_6$  and/or  $P_7$  must then be readjusted. Note that these controls give an instinctive 'wrong' feel: turning them clockwise *reduces* the level.

If the surround speakers give exaggerated reverberation, reduce the delay with  $P_1$ . In the average living room, a delay of 25 ms appears correct: this corresponds to  $P_1$  being almost at its maximum setting.

Do not be surprised if after watching and listening to a number of films, some readjustment of the controls are found desirable.

#### Characteristics

The curves in Fig. 7 and 8 show the amplitude vs frequency characteristics of the centre channel and surround channel respectively. The curves were obtained with an audio analyser. It is evident that they correspond closely with the descriptions. The -3 dB point of the centre channel is at 30 Hz. The -3 dB points of the surround channel are at 50 Hz and 6 kHz; the -6 dB bandwidth is roughly 30 Hz to 7 kHz.

Figure 8 shows that it does not make sense to use tweeters with a linear characteristic up to 20 kHz for the surround channels.

Both curves make it clear that a subwoofer must be connected to the main channels and not to the centre channel or surround channel.

#### Parts list

**Resistors:**  $R_1, R_2, R_{27}, R_{28}, R_{45}, R_{50}, R_{53}-R_{56} = 47 \text{ k}\Omega$  $R_3$ ,  $R_4 = 1 k\Omega$  (see text)  $R_{5}-R_{8}, R_{11}, R_{23} = 10.0 \text{ k}\Omega, 1\%$  $R_9$ ,  $R_{21}$  = 11.8 kΩ, 1%  $R_{10}, R_{22} = 11.0 \text{ k}\Omega, 1\%$  $R_{12}, R_{24} = 6.98 \text{ k}\Omega, 1\%$  $R_{13},\,R_{25}=\,4.32\;k\Omega,\,1\%$  $R_{14}, R_{26} = 5.62 \text{ k}\Omega, 1\%$  $R_{15}=\,18\,k\Omega$  $R_{16}=\,22\;k\Omega$  $R_{17}, R_{18}, R_{36} = 100 \text{ k}\Omega$  $R_{19},\,R_{20}=\,5.6\,k\Omega$  $R_{29}, R_{31} = 2.2 \text{ k}\Omega$  $R_{30}, R_{32} = 6.8 \text{ k}\Omega$  $R_{33}$ ,  $R_{34}$ ,  $R_{37}$ ,  $R_{42}$ ,  $R_{57}$ ,  $R_{59}$ ,  $R_{61}$ ,  $R_{63} = 10 \text{ k}\Omega$  $R_{35} = 100 \Omega$  $R_{38}=\,2.7~k\Omega$  $R_{39}=270\,k\Omega$  $R_{40},\,R_{41}=\,560\;k\Omega$  $R_{43} = 15 k\Omega$  $R_{44} = 12 \text{ k}\Omega$  $R_{46}, R_{48} = 39 \, k\Omega$  $R_{47}, R_{49}, R_{51}, R_{52} = 1 \ k\Omega$  $R_{58}, R_{62} = 560 \Omega$  $R_{60}, R_{64} = 4.7 \Omega$  $R_{65}=\,5.6~M\Omega$  $R_{66}^{00} = 10 \Omega$  $P_1 = 50 \, k\Omega$  preset  $P_2-P_7 = 10 \text{ k}\Omega \text{ preset}$ 

#### Capacitors:

 $\begin{array}{l} C_1, C_2, C_{25}, C_{27}, C_{45}, C_{52} = 2.2 \ \mu\text{F}, \\ \text{polypropylene, pitch 5 mm} \\ C_3, C_5, C_{16}, C_{18} = 2.2 \ n\text{F} \\ C_4, C_{17} = 3.9 \ n\text{F} \\ C_6, C_{19} = 1 \ n\text{F} \\ C_7, C_8, C_{20}, C_{21}, C_{30} - C_{34}, C_{39}, C_{44}, C_{47}, \end{array}$ 

 $C_{10}, C_{11}, C_{13}, C_{35}, C_{63}, C_{64}, C_{67}-C_{70} = 100 \text{ nF},$ ceramic  $C_{12}$ ,  $C_{36}$ ,  $C_{60} = 47 \,\mu\text{F}$ , 25 V, radial  $C_{14}, C_{26} = 4.7 \, \mu F$ , 63 V, radial  $C_{15} = 270 \, pF$  $C_{22} = 68 \, pF$  $C_{23},\,C_{59},\,C_{61},\,C_{62}=\,10~\mu\text{F}\!,\,63$  V, radial  $C_{24} = 1 \ \mu\text{F}$ , polypropylene, pitch 5 mm  $C_{28} = 1 \,\mu\text{F}, 63 \,\text{V}, \text{radial}$  $C_{29} = 220 \,\mu\text{F}, 10 \,\text{V}, \text{ radial}$  $C_{37}$ ,  $C_{38} = 47 \,\mu\text{F}$ , 16 V, radial  $C_{40}, C_{41} = 100 \,\mu\text{F}, 25 \,\text{V}, \text{ radial}$  $C_{42} = 150 \, nF$  $C_{43} = 68 \text{ nF}$  $C_{48}$ ,  $C_{50}$ ,  $C_{55}$ ,  $C_{57} = 220 \,\mu\text{F}$ , 25 V, radial  $C_{65}$ ,  $C_{66} = 4700 \,\mu\text{F}$ , 25 V, radial  $C_{71} = 820 \, nF$ Inductors:  $L_1, L_2 = 100 \,\mu H$  $L_3 = 150 \,\mu H$ Semiconductors:  $D_1-D_4 =$ zener diode, 4.7 V  $D_5 = \bar{1}N4148$  $D_6$  = bi-colour LED (green/red)  $D_8, D_9 = 1N4004$  $B_1 = SB102, 10 A, 100 V$ , for PCB mounting  $T_1 = BC517$ Integrated circuits:  $IC_1, IC_2 = NE5532$  $IC_3 = MN3101$  $IC_4 = MN3008$  $IC_5 = TL072$  $IC_6 = NE571$  $IC_7 = TL084$  $IC_8 = 4030$  $IC_9 = TDA1074A$  $IC_{10}, IC_{11} = TDA2040$  $IC_{12} = 7808$  $IC_{13} = 7908$ Miscellaneous:  $K_1, K_2$  = audio socket for PCB mounting  $K_3 = 3$ -way terminal block, pitch 5 mm  $Re_1 = relay 12 V, 5 A, 270 \Omega$ Heat sink SK57, 37.5 mm high\* Ceramic washers Type AOS220\* Enclosure 300×45×210 mm (W×H×D)  $(11^{7}/_{8} \times 1^{3}/_{4} \times 8^{1}/_{4} \text{ in})$ Mains transformer with 2×12 V, 80 VA secondary Mains entry with integral fuse holder and 500 mA slow fuse Loudspeaker terminals (spring loaded) or suitable audio sockets Mains on/off switch PCB Order No. 950012-1 [950012] \* Available from Dau (UK) Ltd, 70-75 Barnham Road, Barnham PO22 0ES, tele-

phone (01243) 553 031

 $C_{49}, C_{51}, C_{54}, C_{56}, C_{58} = 100 \text{ nF}$ 

 $C_9 = 100 \text{ pF}$  polystyrene, axial

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